

**B.Tech. DEGREE EXAMINATION, JANUARY 2023**

**Thrid Semester**

**Electrical and Electronics Engineering**

**Theory: Electromagnetic Theory**

**(2013-14 Regulations)**

**ANSWER KEY**

**PART A (2X10 = 20 Marks)**

**1. State Divergent theorem (T1)**

The divergence theorem is a mathematical statement of the physical fact that, **in the absence of the creation or destruction of matter, the density within a region of space can change only by having it flow into or away from the region through its boundary.**

$$\iint_S \vec{F} \cdot \hat{n} \, dS = \iiint_V \vec{\nabla} \cdot \vec{F} \, dV$$

**2. Define Energy density (T1)**

Its denoted as  $u_E = \frac{1}{2}\epsilon_0 E^2$  as the energy density, i.e. the energy per unit volume, in the electric field. The energy stored between the plates of the capacitor equals the energy per unit volume stored in the electric field times the volume between the plates.

**3. List the properties of conductor & dielectric materials (T2)**

These are the most important properties of dielectric materials.

1. Electric susceptibility
2. Dielectric polarization
3. Electric dipole moment
4. Electronic polarization

**4. Write Poisson's & Laplace's equation (T1)**

The electric field is related to the charge density by the divergence relationship

$$\vec{\nabla} \cdot \vec{E} = \frac{\rho}{\epsilon_0}$$

$\vec{E}$  = electric field  
 $\rho$  = charge density  
 $\epsilon_0$  = permittivity

The electric field is related to the electric potential by a gradient relationship

$$\vec{E} = -\vec{\nabla} V$$

Therefore the potential is related to the charge density by Poisson's equation

$$\vec{\nabla} \cdot \vec{\nabla} V = \nabla^2 V = \frac{-\rho}{\epsilon_0}$$

In a charge-free region of space, this becomes Laplace's equation

$$\nabla^2 V = 0$$

**5. State Biot Savarts law (model)**

Biot savart law states that “ magnetic field due to a current-carrying conductor at a distance point is inversely proportional to the square of the distance between the conductor and point, and the magnetic field is directly proportional to the length of the conductor, current flowing in the conductor”

Vector notation

$$dB \propto \frac{I \, dl \times r}{r^3}$$

$$r^3$$

### 6. Infer the Lorentz force equation for a moving charge

Lorentz force, the force exerted on a charged particle  $q$  moving with velocity  $v$  through an electric field  $E$  and magnetic field  $B$ . The entire electromagnetic force  $F$  on the charged particle is called the Lorentz force and is given by  $\mathbf{F} = q\mathbf{E} + q\mathbf{v} \times \mathbf{B}$ .

### 7. Find the energy stored in inductor having current of 3A flowing through the inductor of 100mH

Given:  $L = 100\text{mH}$ ;  $I = 3\text{A}$   $W = \frac{1}{2} LI^2$ .  $W = 450 \times 10^{-3} \text{ Joules}$ .

### 8. Identify the relationship between magnetic field intensity & magnetization

The magnetic intensity defines the forces that the poles of a magnet experience in a magnetic field whereas the intensity of magnetization explains the change in the magnetic moment of a magnet per unit volume whereas, Induced magnetization is a process where you can magnetize a non-magnetic material. You can do so when you bring it under the influence of an external magnetic field.

### 9. Write the significance of displacement current

Displacement currents play a central role in the propagation of electromagnetic radiation, such as light and radio waves, through empty space. A traveling, varying magnetic field is everywhere associated with a periodically changing electric field that may be conceived in terms of a displacement current.

In electromagnetism, displacement current is a quantity appearing in Maxwell's equations that is defined in terms of the rate of change of electric displacement field. Displacement current has the same units as electric current, and it is a source of the magnetic field just as actual current is.

### 10. Express the values of skin depth for a plane wave propagation through the dielectric with attenuation constant of $0.2887\text{Np/m}$ (Nepers per meter)

#### CONCEPT:

- The skin effect is the tendency of an **alternating electric current to become distributed within a conductor** such that the current density is largest near the surface of the conductor and decreases with greater depths in the conductor.
- The electric current flows mainly at the skin of the conductor, between the outer surface and a level called the **skin depth**.
- Skin depth is defined as the **reciprocal of attenuation constant** i.e.,

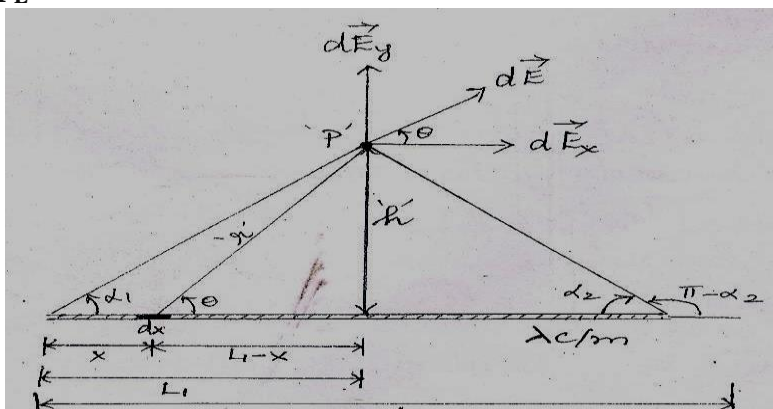
$$\delta = \frac{1}{\alpha}$$

$$= 1/0.2887 = 3.46\text{m}$$

### PART B (5X11 = 55 Marks)

#### UNIT-I

### 11. Obtain the formula for the electric field intensity of an infinite long straight line carrying uniform line charge density of $P_L$



**Assumptions:**

Let the point 'p' be at a distance of 'L<sub>1</sub>' meter from one end of the charged wire along 'x' axis, and at a height 'h' meter along 'y' axis.

Consider a small element 'dx' at a distance 'x' from one end of wire.

Charge element dq=λdx

Total charge 'q' due to full length will be the integration of dq.

Let 'dE' be the field due to charge element λdx

$$\vec{dE} = dE_x \vec{a}_x + dE_y \vec{a}_y \dots \dots \dots (1)$$

From figure;

$$dE_x = dE \cos \theta \vec{a}_x$$

$$dE_y = dE \sin \theta \vec{a}_y$$

By definition;

$$dE = \frac{dq}{4\pi\epsilon_0 r^2} \vec{a}_r \quad \text{V/m}$$

$$\Rightarrow dE_x = \frac{\lambda dx}{4\pi\epsilon_0 r^2} \cos \theta$$

$$\vec{E}_x = \int_{x=0}^L \frac{\lambda dx}{4\pi\epsilon_0 r^2} \cos \theta \vec{a}_x \dots \dots \dots (2)$$

$$\text{From the fig, } \tan \theta = \frac{h}{L_1 - x} \Rightarrow \cot \theta = \frac{L_1 - x}{h}$$

$$\Rightarrow L_1 - x = h \cot \theta$$

$$\Rightarrow -dx = -h \operatorname{cosec}^2 \theta d\theta \dots \dots \dots (3)$$

$$\text{Also, } \sin \theta = \frac{h}{r} \Rightarrow r = h \operatorname{cosec} \theta \dots \dots \dots (4)$$

Sub 3 and 4 in equation 2

$$\vec{E}_x = \int_0^L \frac{\lambda (h \operatorname{cosec}^2 \theta d\theta)}{4\pi\epsilon_0 (h \operatorname{cosec} \theta)^2} \cdot \cos \theta \vec{a}_x$$

NOTE: [ dx → become dθ, so change limits also ie, x=0 to L, becomes θ=α<sub>1</sub>, to π - α<sub>2</sub> ]

$$\vec{E}_x = \int_{\alpha_1}^{\pi - \alpha_2} \frac{\lambda}{4\pi\epsilon_0 h} \cdot \cos \theta d\theta$$

$$\Rightarrow \vec{E}_x = \frac{\lambda}{4\pi\epsilon_0 h} [\sin \theta]_{\alpha_1}^{\pi - \alpha_2} = \frac{\lambda}{4\pi\epsilon_0 h} [\sin (\pi - \alpha_2) - \sin \alpha_1]$$

$$\vec{E}_x = \frac{\lambda}{4\pi\epsilon_0 h} [\sin \alpha_2 - \sin \alpha_1] \quad \text{V/m}$$

Similarly

$$d\vec{E}_y = \frac{\lambda dx}{4\pi\epsilon_0 r^2} \sin\theta \vec{a}_y \quad \Rightarrow \vec{E}_y = \int_{x=0}^L \frac{\lambda dx}{4\pi\epsilon_0 r^2} \sin\theta$$

Substituting for 'θ' and 'dx'

$$\vec{E}_y = \int_{\alpha_1}^{\alpha_2} \frac{\lambda h \cos \theta d\theta}{4\pi\epsilon_0 (h^2 \cos^2 \theta)} \sin\theta \vec{a}_y = \int_{\alpha_1}^{\alpha_2} \frac{\lambda}{4\pi\epsilon_0 h} \sin\theta d\theta \vec{a}_y = \frac{\lambda}{4\pi\epsilon_0 h} [1 - \cos\theta]_{\alpha_1}^{\alpha_2} \vec{a}_y$$

$$\Rightarrow \vec{E}_y = \frac{\lambda}{4\pi\epsilon_0 h} [\cos\alpha_1 + \cos\alpha_2] \vec{a}_y \text{ V/m}$$

Case (i)

If point 'p' is along  $\perp$  r bisector of wire, then  $\alpha_1 = \alpha_2 = \alpha \therefore \vec{a}_x$  component get cancelled and  $\vec{a}_y$  component adds.

$$\vec{E}_y = \vec{E} = \frac{\lambda}{4\pi\epsilon_0 h} [\cos\alpha + \cos\alpha]$$

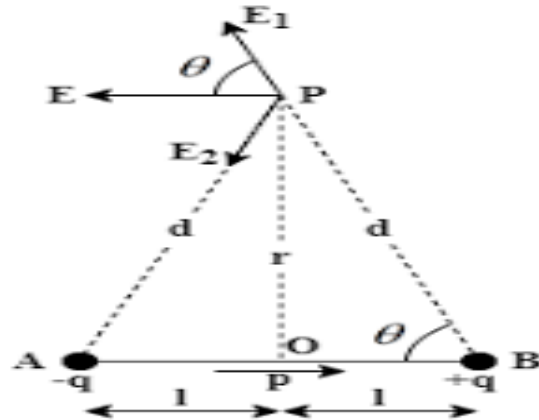
$$\vec{E} = \frac{\lambda}{2\pi\epsilon_0 h} \cdot \cos\alpha \vec{a}_y \text{ V/m}$$

Case (ii)

In case of infinitely long conductor,  $\alpha = 0$

$$\vec{E} = \frac{\lambda}{2\pi\epsilon_0 h} \vec{a}_y \text{ v/m}$$

12. Derive the expression for potential due to an electric dipole at any point P also find the electric fields intensity and in terms of dipole moment



Let an electric dipole consist of two equal and opposite point charges  $-q$  at A and  $+q$  at B, separated by a small distance  $AB = 2a$ , with centre at O.

The dipole moment,  $p = q \times 2a$

We will calculate potential at any point P, where

$OP = r$  and  $\angle BOP = \theta$

Let  $BP = r_1$  and  $AP = r_2$

Draw AC perpendicular PQ and BD perpendicular PO

In  $\triangle AOC$   $\cos\theta = OC/OA = OC/a$

$OC = a\cos\theta$

Similarly,  $OD = a\cos\theta$

Potential at P due to  $+q = \frac{1}{4\pi\epsilon_0} \frac{q}{r_2}$

And Potential at P due to  $-q = \frac{1}{4\pi\epsilon_0} \frac{q}{r_1}$

Net potential at P due to the dipole

$$V = \frac{1}{4\pi\epsilon_0} \left( \frac{q}{r_2} - \frac{q}{r_1} \right)$$

$$\Rightarrow V = \frac{q}{4\pi\epsilon_0} \left( \frac{1}{r_2} - \frac{1}{r_1} \right)$$

Now,  $r_1 = AP = CP$

$= OP + OC$

$= r + a\cos\theta$

And  $r_2 = BP = DP$

$= OP - OD$

$= r - a\cos\theta$

$$V = \frac{q}{4\pi\epsilon_0} \left( \frac{1}{r - a\cos\theta} - \frac{1}{r + a\cos\theta} \right)$$

$$= \frac{q}{4\pi\epsilon_0} \left( \frac{2a\cos\theta}{r^2 - a^2\cos^2\theta} \right)$$

$$= \frac{p\cos\theta}{r^2 - a^2\cos^2\theta} \text{ (Since } p = 2aq \text{)}$$

### Special cases:-

(i) When the point P lies on the axial line of the dipole,  $\theta = 0^\circ$

$$\cos\theta = 1$$

$$V = \frac{p}{r^2 - a^2}$$

$$\text{If } a \ll r, V = \frac{p}{r^2}$$

Thus due to an electric dipole, potential,  $V \propto \frac{1}{r^2}$

(ii) When the point P lies on the equatorial line of the dipole,  $\theta = 90^\circ$

$$\cos\theta = 0$$

i.e electric potential due to an electric dipole is zero at every point on the equatorial line of the dipole.

## UNIT II

13. Find the value of capacitance of a capacitor consisting of two parallel metal plates of 30cm\*30cm surface area separately by 5 mm in air. What is the total energy stored by capacitor is charged to a potential difference of 1000 V . What is energy density?

Sol:-

Area of the parallel plate capacitor =  $30 \times 30 \text{ cm}^2 = 900 \text{ cm}^2$

Distance separating of plates =  $d = 5 \text{ mm} = 5 \times 10^{-3} \text{ m}$

Potential difference across the plate =  $V = 1000 \text{ volts}$

The formula for capacitance will be

$$C = \epsilon_0 A/d$$

electrostatic energy stored in capacitor is given by

$$E = \frac{1}{2} CV^2$$

$$= \frac{1}{2} \frac{\epsilon_0 A}{d} \cdot V^2$$

$\epsilon_0 = \text{permittivity of free space} = 8.854 \times 10^{-12} \text{ C}^2 \text{ N}^{-1} \text{ m}^{-2}$

$$E = \frac{1}{2} \times \frac{8.854 \times 10^{-12} \times 900 \times 10^{-4}}{5 \times 10^{-3}} \times 1000^2$$

$$= 7.9687 \times 10^{-12} \times 10^{-4} \times 10^3 \times 10^6 \times 10^2$$

$$= 7.9687 \times 10^{-16} \times 10^{11} = 7.9687 \times 10^{-5} \text{ J}$$

$\therefore$  The stored electrostatic energy in the capacitor is  $7.9687 \times 10^{-5} \text{ J}$

Energy density  $U = \frac{E}{V} = \frac{E}{A \times d} = \frac{7.9687 \times 10^{-5}}{900 \times 10^{-3} \times 5 \times 10^{-3}}$

$$= \frac{7.9687 \times 10^{-5}}{4500 \times 10^{-6}} = 0.00177 \times 10^{-5} \times 10^6$$

$$= 0.0177 \text{ J/m}^3$$

14. Solve 1D Laplace eq to obtain the fields inside a parallel plate capacitor & also evaluate the expression for the surface charge density at two plates

Electric field on the inside of the parallel plate capacitor due to one of the plates is given by:

$$E_1 = \frac{\sigma}{2\epsilon_0} = \frac{Q}{2A\epsilon_0}$$

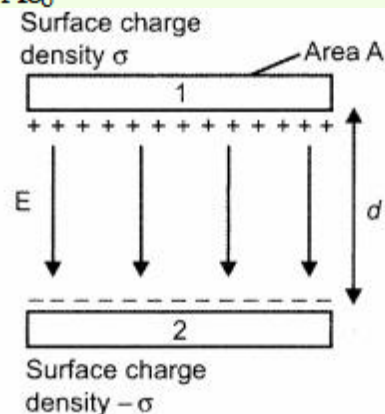
Total electric field due to the two plates is given by:

$$E = 2E_1 = \frac{Q}{A\epsilon_0}$$

Potential difference between the plates is given by:

$$V = Ed$$

$$V = \frac{Qd}{A\epsilon_0}$$



A parallel plate capacitor consists of two large plane parallel conducting plates separated by a small distance. The two plates have charges  $q$  and  $-q$  and distance between them is  $d$ .

Plate 1 has charge density,  $\sigma = \frac{q}{A}$

Plate 2 has charge density,  $\sigma = -\frac{q}{A}$

In the inner region between the plates 1 and 2, the electric fields due to the two charged plates add up.

$$E = \frac{\sigma}{2\epsilon_0} + \frac{\sigma}{2\epsilon_0} = \frac{\sigma}{\epsilon_0} = \frac{q}{A\epsilon_0}$$

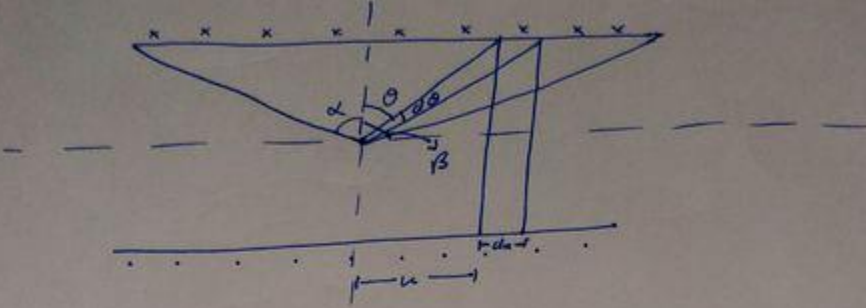


### UNIT III

15. State & explain the Ampere's circuit law & show that the field strength at the end of a solenoid is one half of that at the center

General Solenoid

$R, l$   
 $n \rightarrow$  no. of turns per unit length



$u = R \tan \theta$        $\frac{du}{d\theta} = R \sec^2 \theta$

$dB = \frac{\mu_0 n I R^2}{2(u^2 + R^2)^{3/2}} du$        $B = \int dB = \frac{\mu_0 n I R^2}{2} \int \frac{du}{(u^2 + R^2)^{3/2}}$

$B = \frac{\mu_0 n I R^2 \cdot R}{2 R^2} \int \frac{\sec^2 \theta d\theta}{\sec^3 \theta} = \frac{\mu_0 n I}{2} \int \cos \theta d\theta$

$B = \frac{\mu_0 n I}{2} (\sin \beta + \sin \alpha)$

$\therefore B \text{ at middle for Ideal Solenoid} = \frac{\mu_0 n I \cdot 2 \sin \alpha}{2} = \mu_0 n I$

$\therefore B \text{ at one end for Ideal Solenoid} = \frac{\mu_0 n I \cdot \sin \alpha}{2} = \frac{\mu_0 n I}{2}$



## 16. Derive magnetic scalar potential & vector potential with necessary eq.

### MAGNETIC SCALAR AND VECTOR POTENTIALS

We recall that some electrostatic field problems were simplified by relating the electric Potential  $V$  to the electric field intensity  $\mathbf{E}$  ( $\mathbf{E} = -\text{del } V$ ). Similarly, we can define a potential associated with magnetostatic field  $\mathbf{B}$ . In fact, the magnetic potential could be scalar  $V_m$  vector  $\mathbf{A}$ . To define  $V_m$  and  $\mathbf{A}$  involves two important identities:

$$\text{Del}^*(\text{del } V) = 0$$

$$\text{Del}(\text{del } \times \mathbf{A}) = 0$$

which must always hold for any scalar field  $V$  and vector field  $\mathbf{A}$ .

Just as  $\mathbf{E} = -\text{del } V$ , we define the *magnetic scalar potential*  $V_m$  (in amperes) as related to  $\mathbf{H}$  according to

$$\mathbf{H} = -\text{del } V_m \text{ if } \mathbf{J} = 0$$

$$\mathbf{J} = \text{del } \times \mathbf{H} = -\text{del } \times (-\text{del } V_m) = 0$$

since  $V_m$ , must satisfy the condition in eq. Thus the magnetic scalar potential  $V_m$  is only defined in a region where  $\mathbf{J} = 0$  as in eq. We should also note that  $V_m$  satisfies Laplace's equation just as  $V$  does for electrostatic fields; hence,

$$\text{del }^2 V_m = 0, (\mathbf{J} = 0)$$

### SCALAR MAGNETIC POTENTIAL

Like scalar electrostatic potential, it is possible to have scalar magnetic potential. It is defined in such a way that its negative gradient gives the magnetic field, that is,

$$\mathbf{H} = \text{del } V_m$$

Taking curl on both sides, we get

$$\text{del } \times \mathbf{H} = -\text{del } \times \text{del } V_m$$

But curl of the gradient of any scalar is always zero.

$$\text{So, } \text{del } \times \mathbf{H} = 0$$

But, by Ampere's circuit law  $\text{del}^* \mathbf{H} = \mathbf{J}$

$$\text{or, } \mathbf{J} = 0$$

In other words, scalar magnetic potential exists in a region where  $\mathbf{J} = 0$ .

$$\mathbf{H} = -\text{del } V_m (\mathbf{J}=0)$$

The scalar potential satisfies Laplace's equation, that is, we have

$$\text{del} \cdot \mathbf{B} = \mu_0 \text{del} \cdot \mathbf{H} = 0 = \text{del} \cdot (-\text{del } V_m) = 0$$

or,

$$\text{del }^2 V_m = 0 (\mathbf{J} = 0)$$

#### Characteristics of Scalar Magnetic Potential ( $V_m$ )

1. The negative gradient of  $V_m$  gives  $\mathbf{H}$ , or  $\mathbf{H} = -\nabla V_m$
2. It exists where  $\mathbf{J} = 0$
3. It satisfies Laplace's equation.

$$V_m = -\int_A^B \mathbf{H} \cdot d\mathbf{L}$$

4. It is directly defined as
5. It has the unit of Ampere.

### VECTOR MAGNETIC POTENTIAL

**Vector magnetic potential** exists in regions where  $\mathbf{J}$  is present. It is defined in such a way that its curl gives the magnetic flux density, that is,

$$\mathbf{B} = \text{del } \times \mathbf{A}$$

where  $\mathbf{A}$  = vector magnetic potential (wb/m).

It is also defined as

$$A \equiv \oint \frac{\mu_0 I dL}{4\pi R} \left( \frac{\text{Henry} - \text{Amp}}{m} \right)$$

or, 
$$A \equiv \oint_s \frac{\mu_0 K ds}{4\pi R}, \quad (K = \text{current sheet})$$

or, 
$$A \equiv \oint_v \frac{\mu_0 J dv}{4\pi R},$$

### Characteristics of Vector Magnetic Potential

1. It exists even when  $\mathbf{J}$  is present.
2. It is defined in two ways

$$\mathbf{B} \equiv \nabla \times \mathbf{A} \quad \text{and}$$

$$\oint_v \frac{\mu_0 J dv}{4\pi R}$$

3.  $\nabla^2 \mathbf{A} = \mu_0 \mathbf{j}$
4.  $\nabla^2 \mathbf{A} = 0$  if  $\mathbf{J} = 0$

## UNIT IV

17. A solenoid is 50 cm long 2 cm in dia & 1500 turns. The cylindrical core has a diameter of 2 cm & relative permeability of 75. This is co-axial with 2<sup>nd</sup> which is 50 cm long 3 cm dia 1200 turns. Solve the inductance  $L$  for inner & outer solenoid.

Given

length of the solenoid,  $l_1 = 50 \text{ cm} = 0.5 \text{ m}$

diameter,  $d_1 = 2 \text{ cm} \Rightarrow R_1 = 0.01 \text{ m}$

Number of turns,  $N_1 = 1500$

cylinder core diameter,  $d_c = 2 \text{ cm} \Rightarrow R_c = 0.01 \text{ m}$

relative permeability,  $\mu_r = 75$

$l_2 = 50 \text{ cm} = 0.5 \text{ m}$ ,  $d_2 = 3 \text{ cm} \Rightarrow 1.5 \text{ cm} = R_2$ ,  $N_2 = 1200$

$L_1 = ?$  and  $L_2 = ?$

$$L_1 = \frac{\mu_1 N_1^2 S_1}{l} = \frac{\mu_r \mu_0 N_1^2 \pi D_1^2}{4l}$$

$$= \frac{(75)(4\pi \times 10^{-7} \text{ H/m})(1500)^2 \pi (.02 \text{ m})^2}{4(.50 \text{ m})}$$

$$L_1 = .133 \text{ H} = 133 \text{ mH}$$

Similarly

$$L_2 = \frac{\mu_1 N_1^2 S_1}{l} = \frac{\mu_r \mu_0 N_1^2 \pi D_1^2}{4l}$$

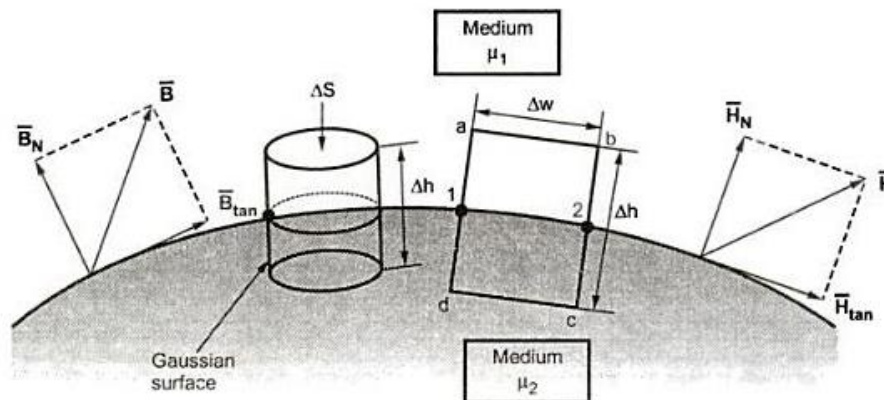
$$= \frac{(75)(4\pi \times 10^{-7} \text{ H/m})(1200)^2 \pi (.03 \text{ m})^2}{4(.50 \text{ m})}$$

$$L_2 = 192 \text{ mH}$$

**18. Describe about the magnetic boundary condition at the interface between two magnetic medium & derive the necessary boundary condition**

The conditions of magnetic field existing at the boundary of the two media when the magnetic field passes from one medium to other are called boundary conditions or magnetic boundary conditions. In the magnetic boundary condition, the condition of B and H are studied at the boundary. B and H vectors are resolved into two components.

1. **Tangential to boundary**
2. **Normal (perpendicular) to boundary**



Consider a boundary between two isotropic homogeneous linear materials with different permeabilities  $\mu_1$  and  $\mu_2$ .

Boundary conditions for normal component

To find the normal component of E select a closed Gaussian surface in the form of a right circular cylinder.

Let the height be  $\Delta h$  and be placed in such a way the  $\Delta h / 2$  is in medium 1 and remaining  $\Delta h / 2$  is in medium 2.

Also z-axis of the cylinder is in the normal direction to the surface.

According to Gauss's law for the magnetic field

$$\oint_S \vec{B} \cdot d\vec{S} = 0$$

$$\oint_{\text{top}} \vec{B} \cdot d\vec{S} + \oint_{\text{bottom}} \vec{B} \cdot d\vec{S} + \oint_{\text{lateral}} \vec{B} \cdot d\vec{S} = 0$$

For top surfaces :

$$\oint_{\text{Top}} \vec{B} \cdot d\vec{S} = B_{N1} \oint_{\text{Top}} d\vec{S} = B_{N1} \Delta S$$

For bottom surface :

$$\oint_{\text{Bottom}} \vec{B} \cdot d\vec{S} = B_{N2} \oint_{\text{Bottom}} d\vec{S} = B_{N2} \Delta S$$

For lateral surface

$$\oint_{\text{Lateral}} \vec{B} \cdot d\vec{S} = 0$$

Sub the above equations

$$B_{N1} \Delta S - B_{N2} \Delta S = 0$$

$B_{N1}$  and  $B_{N2}$  are in opposite direction

$$B_{N1} \Delta S = B_{N2} \Delta S$$

$$\boxed{B_{N1} = B_{N2}}$$

Thus the normal component of  $B$  is continuous at the boundary. As the magnetic flux density and the magnetic field intensity are related by

$$\vec{B} = \mu \vec{H}$$

The above equation becomes

$$\mu_1 H_{N1} = \mu_2 H_{N2}$$

$$\boxed{\frac{H_{N1}}{H_{N2}} = \frac{\mu_2}{\mu_1} = \frac{\mu_{r2}}{\mu_{r1}}}$$

Hence the normal components of  $H$  is not continuous at the boundary. The field strengths in two media are inversely proportional to their relative permeabilities.

### Boundary conditions for tangential component

According to Ampere's circuital law

$$\oint \vec{H} \cdot d\vec{L} = I$$

Consider a rectangular loop in clockwise direction as a-b-c-d-a. This closed path is placed in a plane normal to the boundary surface

$$\oint \vec{H} \cdot d\vec{L} = \int_a^b \vec{H} \cdot d\vec{L} + \int_b^c \vec{H} \cdot d\vec{L} + \int_c^d \vec{H} \cdot d\vec{L} + \int_d^a \vec{H} \cdot d\vec{L} = I$$

The sides a-b and c-d are parallel to the tangential direction to the surface. While the other two sides are normal to the surface at the boundary

Let the elementary height and elementary width be  $\Delta h$  and  $\Delta w$

The normal and tangential component in medium 1 and in medium 2 are in opposite direction

The above equation are written as

To get conditions at boundary,  $\Delta h \rightarrow 0$ . Thus,

$$K \cdot dw = H_{tan1}(\Delta w) - H_{tan2}(\Delta w)$$

$$H_{tan1} - H_{tan2} = K$$

In vector form

$$\vec{H}_{tan1} - \vec{H}_{tan2} = \vec{a}_{N12} \times \vec{K}$$

$\vec{a}_{N12}$  – unit vector

the above equ in terms of B

$$\frac{B_{tan1}}{\mu_1} - \frac{B_{tan2}}{\mu_2} = K$$

Special case

The boundary is free of current

So  $K = 0$

$$H_{tan1} - H_{tan2} = 0$$

$$H_{tan1} = H_{tan2}$$

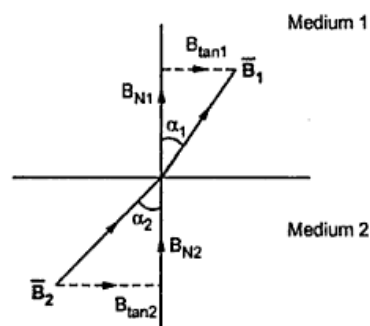
Similarly

$$\frac{B_{tan1}}{\mu_1} = \frac{B_{tan2}}{\mu_2}$$

$$\frac{B_{tan1}}{B_{tan2}} = \frac{\mu_1}{\mu_2} = \frac{\mu_{r1}}{\mu_{r2}}$$

Condition

From the above equ the tangential components of H are continuous while the tangential components of B are discontinuous at the boundary



In medium 1

$$\tan \alpha_1 = B_{tan1} / B_{N1}$$

In medium 2

$$\tan \alpha_2 = B_{tan2} / B_{N2}$$

Dividing the above equ

$$\tan \alpha_1 / \tan \alpha_2 = B_{tan1} B_{N2} / B_{tan2} B_{N1}$$

W.K.T

$$B_{N1} = B_{N2}$$

$$\tan \alpha_1 / \tan \alpha_2 = B_{tan1} / B_{tan2} = \mu_{r1} / \mu_{r2}$$

## UNIT V

### 19.(a) State & prove Poynting Thorem

Poynting's theorem states that the rate of energy transfer per unit volume from a region of space equals the rate of work done on the charge distribution in the region, plus the energy flux leaving that region.

**Proof :** The energy density carried by the electromagnetic wave can be calculated using Maxwell's equations

$$\text{as } \operatorname{div} \vec{D} = 0 \quad \dots(i) \quad \operatorname{div} \vec{B} = 0 \quad \dots(ii) \quad \operatorname{Curl} \vec{E} = -\frac{\partial \vec{B}}{\partial t} \quad \dots(iii)$$

$$\text{and } \operatorname{Curl} \vec{H} = \vec{J} + \frac{\partial \vec{D}}{\partial t} \quad \dots(iv)$$

taking scalar product of (iii) with  $\vec{H}$  and (iv) with  $\vec{E}$

$$\text{i.e.} \quad \vec{H} \operatorname{curl} \vec{E} = -\vec{H} \cdot \frac{\partial \vec{B}}{\partial t} \quad \dots(v)$$

$$\text{and} \quad \vec{E} \cdot \operatorname{curl} \vec{H} = \vec{E} \cdot \vec{J} + \vec{E} \cdot \frac{\partial \vec{D}}{\partial t} \quad \dots(vi)$$

$$\begin{aligned} \text{doing (vi) - (v) i.e.} \quad & \vec{H} \cdot \operatorname{curl} \vec{E} - \vec{E} \cdot \operatorname{curl} \vec{H} = -\vec{H} \cdot \frac{\partial \vec{B}}{\partial t} - \vec{E} \cdot \vec{J} - \vec{E} \cdot \frac{\partial \vec{D}}{\partial t} \\ & = -\left[ \vec{H} \cdot \frac{\partial \vec{B}}{\partial t} + \vec{E} \cdot \frac{\partial \vec{D}}{\partial t} \right] - \vec{E} \cdot \vec{J} \end{aligned}$$

$$\text{as} \quad \operatorname{div} (\vec{A} \times \vec{B}) = \vec{B} \cdot \operatorname{curl} \vec{A} - \vec{A} \cdot \operatorname{curl} \vec{B}$$

$$\text{so} \quad \operatorname{div} (\vec{E} \times \vec{H}) = -\left[ \vec{H} \cdot \frac{\partial \vec{B}}{\partial t} + \vec{E} \cdot \frac{\partial \vec{D}}{\partial t} \right] - \vec{E} \cdot \vec{J} \quad \dots(vii)$$

$$\text{But} \quad \vec{B} = \mu \vec{H} \quad \text{and} \quad \vec{D} = \epsilon \vec{E}$$

$$\begin{aligned} \text{so} \quad \vec{H} \cdot \frac{\partial \vec{B}}{\partial t} &= \vec{H} \cdot \frac{\partial}{\partial t} (\mu \vec{H}) = \frac{1}{2} \mu \frac{\partial}{\partial t} (H^2) \\ &= \frac{\partial}{\partial t} \left[ \frac{1}{2} \vec{H} \cdot \vec{B} \right] \end{aligned}$$

$$\text{and} \quad \vec{E} \cdot \frac{\partial \vec{D}}{\partial t} = \vec{E} \cdot \frac{\partial}{\partial t} (\epsilon \vec{E}) = \frac{1}{2} \epsilon \frac{\partial}{\partial t} (E)^2 = \frac{\partial}{\partial t} \left[ \frac{1}{2} \vec{E} \cdot \vec{D} \right]$$

$$\text{so from equation (vii)} \quad \operatorname{div} (\vec{E} \times \vec{H}) = -\frac{\partial}{\partial t} \left[ \frac{1}{2} (\vec{H} \cdot \vec{B} + \vec{E} \cdot \vec{D}) \right] - \vec{E} \cdot \vec{J}$$

$$\text{or} \quad \vec{E} \cdot \vec{J} = -\frac{\partial}{\partial t} \left[ \frac{1}{2} (\vec{H} \cdot \vec{B} + \vec{E} \cdot \vec{D}) \right] - \operatorname{div} (\vec{E} \times \vec{H}) \quad \dots(viii)$$

Integrating equation (viii) over a volume V enclosed by a surface S

$$\int_V \vec{E} \cdot \vec{J} dV = -\int_V \frac{\partial}{\partial t} \left\{ \frac{1}{2} (\vec{H} \cdot \vec{B} + \vec{E} \cdot \vec{D}) \right\} dV - \int_V \operatorname{div} (\vec{E} \times \vec{H}) dV$$



$$\begin{aligned}
\text{or} \quad \int_V \vec{E} \cdot \vec{J} \, dV &= - \int_V \left[ \frac{1}{2} \mu H^2 + \frac{1}{2} \epsilon E^2 \right] dV - \int_S (\vec{E} \times \vec{H}) \cdot d\vec{s} \\
\text{as} \quad \vec{B} &= \mu \vec{H}, \quad \vec{D} = \epsilon \vec{E} \quad \text{and} \quad \int_V \text{div}(\vec{E} \times \vec{H}) dV = \int_S (\vec{E} \times \vec{H}) \cdot d\vec{s} \\
\text{or} \quad \int_V (\vec{E} \cdot \vec{J}) dV &= - \frac{\partial}{\partial t} \int_V \left[ \frac{1}{2} \mu H^2 + \frac{1}{2} \epsilon E^2 \right] dV - \int_S (\vec{E} \times \vec{H}) \cdot d\vec{s} \\
\text{or} \quad \int_S (\vec{E} \times \vec{H}) \cdot d\vec{s} &= - \int_V \frac{\partial U_{em}}{\partial t} dV - \int_V (\vec{E} \cdot \vec{J}) dV \\
\text{or} \quad \boxed{\int_S \vec{P} \cdot d\vec{s} = - \int_V \frac{\partial U_{em}}{\partial t} dV - \int_V (\vec{E} \cdot \vec{J}) dV} &\quad \left( \text{as } \vec{P} = \vec{E} \times \vec{H} \right) \quad \dots(\text{ix})
\end{aligned}$$

i.e. Total power leaving the volume = rate of decrease of stored e.m. energy -  
ohmic power dissipated due to charge motion

This equation (ix) represents the poynting theorem according to which the net power flowing out of a given volume is equal to the rate of decrease of stored electromagnetic energy in that volume minus the conduction losses.

(b) Describe the Poynting Vector , Pag & Instantaneous Power.  
For Answer Refer 19(a)

20. In a material for which  $\sigma=5\text{S/m}$  &  $\epsilon_r=1$ , the electric field intensity is  $E = 250 \sin 10^{10}t \text{ v/m}$ . Estimate the conduction & displacement current densities, & the frequency at which both have equal magnitudes

Given that,

$$\sigma = 5 \text{ S/m}$$

Electric field intensity is

$$E = 250 \sin 10^{10}t$$

We need to calculate the conductor current

Using formula of conductor current

$$I_c = \sigma E$$

Put the value into the formula

$$I_c = 5 \times 250 \sin(10^{10}t)$$

$$I_c = 1250 \sin(10^{10}t) \text{ C/m}^2$$

We need to calculate the displacement current densities

Using formula of displacement current density

$$I_d = \epsilon_0 \frac{dE}{dt}$$

Put the value into the formula

$$I_d = 8.854 \times 10^{-12} \times 250 \times 10^{10} \cos(10^{10}t)$$

$$I_d = 22.14 \cos(10^{10}t) \text{ C/m}^2$$

The frequency at which they have equal magnitude.

$$I_c = I_d$$

$$\sigma = \omega \epsilon_0$$

Then,

We need to calculate the frequency at which they have equal magnitude

Using formula of frequency

$$\omega = \frac{\sigma}{\epsilon_0}$$

Put the value into the formula

$$\omega = \frac{5}{8.854 \times 10^{-12}}$$

$$\omega = 5.64 \times 10^{11} / \text{sec}$$

$$\omega = 89.8 \text{ GHz}$$

Hence, The conductor current is  $1250 \sin(10^{10}t) \text{ C/m}^2$

The displacement current density is  $22.14 \cos(10^{10}t) \text{ C/m}^2$

The frequency is 89.8 GHz.